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Hirasugar Institute of Technology, Nidasoshi.

Inculcating Values, Promoting Prosperity

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ECE Dept.

Exam.

Internal Assessment

Even Sem(2017-18)

SECOND INTERNAL ASSESSMENT

Sem: VI

Date: 11/04/2018

Sub: Digital Communication

Time: 11:00am-12:00pm

Sub. Code: 15EC61

Max. Marks: 25

Note: Answer two full questions, draw sketches wherever necessary.

Q. No		Description of Question	Marks	CO	RBT LEVEL
1	a	Explain with block diagram QPSK transmitter and Receiver.	6	C301.3	L1,L2,L3
	b	Obtain the expression for average probability of error calculation for FSK with coherent receiver.	7	C301.3	L1,L2,L3
OR					
2	a	Explain Gram Schmidt Orthogonalization procedure.	6	C301.2	L1,L2,L3
	b	Consider the set of signals $S_i(t) = \begin{cases} \sqrt{2E/T} \cos(2\pi f_c t - i\pi/4) & 0 \leq t \leq T \\ 0 & \text{elsewhere} \end{cases}$ where $i=0,1,2,3$ and f_c is an integer multiple of $1/T$. i) Determine a set of orthogonality N of the signal set. ii) Determine the coefficients S_{ij} of the signals $S_i(t)$.	7	C301.2	L1,L2,L3
3	a	Explain Non coherent DPSK System.	6	C301.3	L1,L2,L3
	b	Illustrate the operation of DPSK for the binary sequence 10010011 and find the encoded and decoded sequence.	6	C301.3	L1,L2,L3
OR					
4	a	Explain the geometric representation of signals.	6	C301.2	L1,L2,L3
	b	Three signals $s_1(t)$, $s_2(t)$ and $s_3(t)$ are given as follows. Apply the Gram Schmidt procedure to obtain an orthonormal basis for the signals. $s_1(t)=3 ; 0 < t < 4$ $s_2(t)=3 ; 0 < t < 2$ $s_3(t)=3 ; 2 < t < 4$	6	C301.2	L1,L2,L3

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SCHEME OF EVALUATION IA- II

Sem : VI		Subject : Digital Communication	Sub Code : 15EC61	Date : 11/04/2018		
Q. No.	Bit	Description	Marks	CO's	RBT LEVEL	
1)	a)	<p><u>QPSK Transmitter :-</u></p> <p>+ To split a bit, demultiplexers will be used. $b_1(t)$ is odd & $b_2(t)$ is even.</p> <p>* Here $D_1(t) = \sqrt{\frac{2}{T}} \cos 2\pi f_c t$ → 3M</p> <p>$D_2(t) = \sqrt{\frac{2}{T}} \sin 2\pi f_c t$ → 3M</p> <p><u>QPSK Receiver :-</u></p> <p>Received signal $s(t)$</p> <p>$D_1(t)$ In Phase → 3M</p> <p>$D_2(t)$ Quadrature Phase → 3M</p> <p>Decision device</p> <p>Decision device</p> <p>MUX → Binary wave</p> <p>If $a_1 > 1$ → Decision is in favour of '1' If $a_1 < 1$ → Decision is in favour of '0' If $a_2 > 1$ → Decision is in favour of '1' If $a_2 < 1$ → Decision is in favour of '0' * Two O/Ps of combined in multiplexer & reproduce original signal</p>	3M	C301.3	L1, L2, L3	

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SCHEME OF EVALUATION IA-

Sem : VI		Subject : DC	Sub Code : 15EC61	Date : 11/04/2018	Marks	CO's	RBT LEVEL
Q. No.	Bit	Description					
1-)	b)	<p><u>Average Probability of error calculation for FSK with Coherent receiver:-</u></p> <p>Received signal: $s(t) = s_1(t) + w(t)$</p> <p>$E(w) = E[s_1] - E[s_2] = 0 - \sqrt{E_b} = -\sqrt{E_b}$</p> <p>$Var(w) = Var(s_1) + Var(w_2) = N_0 \rightarrow 3M$</p> <p>$f_x(y 0) = \frac{1}{\sqrt{2\pi N_0}} e^{-\frac{[y + \sqrt{E_b}]^2}{2N_0}}$</p> <p>$\therefore$ By using above conditional PDF we can calculate $P_e(w)$</p> <p>$\therefore P_e(w) = \frac{1}{2} \text{erfc} \left[\sqrt{\frac{E_b}{2N_0}} \right]$</p> <p>ii) $P_e(1) = \frac{1}{2} \text{erfc} \left[\sqrt{\frac{E_b}{2N_0}} \right] \rightarrow 3M$</p> <p>$\therefore P_e = P(w)P_e(w) + P(1)P_e(1)$</p> <p>$P_e = \frac{1}{2} \left[\frac{1}{2} \text{erfc} \left[\sqrt{\frac{E_b}{2N_0}} \right] + \frac{1}{2} \text{erfc} \left[\sqrt{\frac{E_b}{2N_0}} \right] \right]$</p> <p>$\boxed{P_e = \frac{1}{2} \text{erfc} \left(\sqrt{\frac{E_b}{2N_0}} \right)} \rightarrow 1M$</p>				C3013 L1, L2, L3	

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SCHEME OF EVALUATION IA- II

Sem : VI		Subject : DC	Sub Code : 102EC 61	Date : 11 / 04 / 2018		
Q. No.	Bit	Description	Marks	CO's	RBT LEVEL	
2)	a)	<p>erran schmidt's orthogonalization procedure consists of two steps.</p> <p><u>Step 1</u> :- we need check given set of the signals are independent or not</p> <p>i.e $a_1 s_1(t) + a_2 s_2(t) + \dots + a_m s_m(t) = 0$</p> <p>$\therefore a_m = 0$</p> <p>$s_m(t) = - \left[\frac{a_1}{a_m} s_1(t) + \frac{a_2}{a_m} s_2(t) + \dots + \frac{a_{m-1}}{a_m} s_{m-1}(t) \right]$</p> <p>Here we need to check the above equation is independent or not if not, there exist b_1, b_2, \dots, b_{m-1} constants. $\rightarrow 2M$</p> <p>$\therefore s_{m-1} = - \left[\frac{b_1}{b_{m-1}} s_1(t) + \frac{b_2}{b_{m-1}} s_2(t) + \dots + \frac{b_{m-2}}{b_{m-1}} s_{m-2}(t) \right]$</p> <p>continuing in this way finally we need to conclude that set of the signals are linearly independent.</p> <p><u>Step 2</u> :- Orthogonal basis functions are having unit energy & they are orthogonal to each other</p> <p>i.e $\phi_j(t) = \frac{g_j(t)}{\sqrt{\int g_j^2(t) dt}}$ $\rightarrow 4M$</p> <p>$g_j(t) = s_j(t) - \sum_{i=1}^{j-1} s_{i,j} \phi_i(t)$</p>		CO-2	L1, L2, L3	

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SCHEME OF EVALUATION IA-

Sem : V /		Subject : DC	Sub Code : 15EC61	Date : 11/04/2018		
Q. No.	Bit	Description	Marks	CO's	RBT LEVEL	
2)	b)	$S_1(t) = \sqrt{\frac{2E}{T}} \cos(2\pi f_c t - i\pi/4)$ $S_0(t) = \sqrt{\frac{2E}{T}} \cos 2\pi f_c t$ $S_1(t) = \sqrt{\frac{2E}{T}} \left[\frac{1}{\sqrt{2}} \cos 2\pi f_c t + \frac{1}{\sqrt{2}} \sin 2\pi f_c t \right]$ $S_2(t) = \sqrt{\frac{2E}{T}} \sin 2\pi f_c t$ $S_3(t) = \sqrt{\frac{2E}{T}} \left[-\frac{1}{\sqrt{2}} \cos 2\pi f_c t + \frac{1}{\sqrt{2}} \sin 2\pi f_c t \right]$ <p>i) $Q_1(t) = \sqrt{\frac{2E}{T}} \cos 2\pi f_c t ; Q_2(t) = \sqrt{\frac{2E}{T}} \sin 2\pi f_c t$ → 2M</p> <p>ii) $S_0 = (\sqrt{E}, 0)$ $S_2 = (\sqrt{E}/2, \sqrt{E}/2)$ S_{01}, S_{02} S_{21}, S_{22}</p> <p>$S_1 = (\sqrt{E}/2, \sqrt{E}/2)$ $S_3 = (0, \sqrt{E})$ S_{11}, S_{12} S_{31}, S_{32}</p> <p>$S_3 = (-\sqrt{E}/2, \sqrt{E}/2)$ → 5M S_{31}, S_{32}</p>		C30.2	L1, L2, L3	
3)	a)	<p><u>DPSK Transmitter :-</u></p> <p><u>DPSK Receiver :-</u></p> <p>Explaining above block diagrams → 3M</p>		C30.3	L1, L2, L3	

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SCHEME OF EVALUATION IA-

Sem : VI		Subject : DC	Sub Code : 15EC 01	Date : 11/04/2018	Marks	CO's	RBT LEVEL
Q. No.	Bit	Description					
3)	b)	<p>Binary Sequence</p> <p>∴ Encoded sequence is 0, 0, 1, 0, 0, 1, 0, 0, 0</p> <p>∴ Decoded sequence is 1, 0, 0, 1, 0, 0, 1, 1</p>			3m 3m	C301.3	L1, L2, L3
4)	a)	<p><u>Orthometric Representation of signals</u></p> <p>Let ϕ be the energy signals</p> $S_i(t) = \{ S_{i1}(t), S_{i2}(t), \dots, S_{in}(t) \}$ $\Phi_j(t) = \{ \Phi_1(t), \Phi_2(t), \dots, \Phi_n(t) \}$ $\therefore S_i(t) = S_{i1}\Phi_1 + S_{i2}\Phi_2 + \dots + S_{in}\Phi_n(t)$ $= \sum_{j=1}^n S_{ij}\Phi_j(t)$ <p>where $S_{ij} = \int_0^T S_i(t)\Phi_j(t) dt$</p> $S_i = \begin{bmatrix} S_{i1} \\ S_{i2} \\ \vdots \\ S_{in} \end{bmatrix} \quad i=1, 2, \dots, m$ <p>This is an example of signal space of $N=2$ & $M=3$</p>			4m 2m	C301.2	L1, L2, L3

SCHEME OF EVALUATION IA-

Sem : VI		Subject : DC	Sub Code : 15EC61	Date : 11/04/2018	Marks	CO's	RBT LEVEL
Q. No.	Bit	Description					
4)	b)	$S_1(t) = 3; 0 \leq t \leq 4$ $S_2(t) = 3; 0 \leq t \leq 2$ $S_3(t) = 3; 2 \leq t \leq 4.$ $\therefore E_1 = \int_0^4 S_1^2(t) dt = 36$ $\therefore \Phi_1(t) = \frac{S_1(t)}{\sqrt{E_1}} = \frac{3}{\sqrt{36}} = \frac{3}{6} = \frac{1}{2}$ $\therefore \Phi_1(t) = \begin{cases} \frac{1}{2} & 0 \leq t \leq 4 \\ 0 & \text{elsewhere} \end{cases} \rightarrow 2M$ <p>WKT</p> $S_{21} = \int_0^4 S_2(t) \Phi_1(t) dt$ $S_{21} = \int_0^2 3 \times \frac{1}{2} = \frac{3}{2} \times 2 = 3$ $\boxed{S_{21} = 3}$ $g_2(t) = S_2(t) - S_{21} \Phi_1(t)$ $\int_0^4 g_2^2(t) dt = \int_0^4 [S_2(t) - 3\Phi_1(t)]^2 dt$ $= 9 \times 2 + 9 \times \frac{1}{4} \times 4 - 2 \times 3 \times 3 \times \frac{1}{2} \times 2$ $= 18 + 9 - 18 = 9.$ $\therefore \sqrt{\int_0^4 g_2^2(t) dt} = 3$ $\therefore \Phi_2(t) = \frac{1}{3} [S_2(t) - 3\Phi_1(t)]$ $\therefore \Phi_2(t) = \begin{cases} \frac{1}{2}, & 0 \leq t \leq 2 \\ -\frac{1}{2} & 2 \leq t \leq 4 \\ 0 & \text{otherwise} \end{cases} \rightarrow 4M$				CO-2	C1, C2, C3