

Department of Electronics & Communication Engg.

Course : Signals and Systems E

Engg-15EC44.

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Finding Unit Impulse Response and step response S(n)

Stability and Causality

Definition: A system is stable if and only if every bounded input produces a bounded output. A bounded input/output is a signal for which for all values of *t*.

Stability for LTI Systems
$$\int_{-\infty}^{\infty} |h(t)| dt < \infty$$

- if and only if holds in either direction

 A theorem which applies to LTI systems states that a system(LTI system) is stable if and only of

Causality

Definition: A system is causal if and only if the output at the present time does not depend upon future values of the input.

Causal for LTI Systems h(t) = 0 for t < 0

• A theorem which applies to LTI systems is

<u>Example</u>: Step Response from $h(t) = e^{-at}u(t)$

- Knowing the impulse response of a system we can find the respond to a step input by just integrating the output, since *u(n)* at the input is obtained by integrating δ(*t*)
- Thus we can write that

$$y(t) = u(t) * h(t) = \int_{-\infty}^{t} h(\tau) d\tau$$
$$= \int_{-\infty}^{t} e^{-a\tau} u(\tau) d\tau = \int_{0}^{t} e^{-a\tau} d\tau$$
$$= \frac{e^{-a\tau}}{-a} \Big|_{0}^{t} = \frac{1}{a} [1 - e^{-at}] u(t)$$

• This result is consistent with earlier analysis

<u>Example</u>: LTI with $h(t) = e^{-at}u(t)$

• For stability

$$\int_{-\infty}^{\infty} \left| e^{-at} u(t) \right| dt = \int_{0}^{\infty} e^{-at} dt$$
$$= \left. \frac{e^{-at}}{-a} \right|_{0}^{\infty} = \frac{1}{a}, a > 0$$

- We must have a > 0 for stability
- Note that a = 0 result in h(t) = u(n), which is an integrator system, hence an integrator system is not stable



The Fourier series of a periodic continuous-time signal

$$x(t) = \sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t} = \sum_{k=-\infty}^{+\infty} a_k e^{jk(2\pi/T)t}$$
$$a_k = \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt = \frac{1}{T} \int_T x(t) e^{-jk(2\pi/T)t} dt$$

Eq. (3.32) is referred to as the Synthesis equation, and Eq. (3.33) is referred to as analysis equation. The set of coefficient { } k a are often called the Fourier series coefficients of the spectral coefficients of x(t).

The coefficient 0 a is the dc or constant component and is given with k = 0, that is

$$a_0 = \frac{1}{T} \int_T x(t) dt$$

,

Properties of the Continuous-Time Fourier Series

Let x(t) and y(t) denote two periodic signals with period T and which have Fourier series coefficients denoted by a_k and b_k , that is

$$x(t) \xleftarrow{FS} a_k \text{ and } y(t) \xleftarrow{FS} b_k,$$

then we have

$$z(t) = Ax(t) + By(t) \xleftarrow{FS} c_k = Aa_k + Bb_k.$$
(3.48)

3.5.2 Time Shifting

When a time shift to a periodic signal x(t), the period T of the signal is preserved.

If $x(t) \xleftarrow{FS}{} a_k$, then we have

$$x(t-t_0) \xleftarrow{FS} e^{-jk\omega_0 t} a_k . \tag{3.49}$$

The magnitudes of its Fourier series coefficients remain unchanged.

3.4.3 Time Reversal

If
$$x(t) \xleftarrow{FS} a_k$$
, then

$$x(-t) \xleftarrow{FS} a_{-k}. \tag{3.50}$$

Time reversal applied to a continuous-time signal results in a time reversal of the corresponding sequence of Fourier series coefficients.

If x(t) is even, that is x(t) = x(-t), the Fourier series coefficients are also even, $a_{-k} = a_k$. Similarly, if x(t) is odd, that is x(-t) = -x(t), the Fourier series coefficients are also odd, $a_{-k} = -a_k$.

3.5.4 Time Scaling

If x(t) has the Fourier series representation $x(t) = \sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t}$, then the Fourier series representation of the time-scaled signal $x(\alpha t)$ is

Parseval's Relation for Continuous-Time periodic Signals is

$$\frac{1}{T}\int_{T}\left|x(t)\right|^{2}dt = \sum_{k=-\infty}^{\infty}\left|a_{k}\right|^{2},$$



Example: The periodic square wave, sketched in the figure below and define over one period is

$$x(t) = \begin{cases} 1, & |t| < T_1 \\ 0, & T_1 < |t| < T/2 \end{cases}$$
(3.35)

The signal has a fundamental period T and fundamental frequency $\omega_0 = 2\pi / T$.



Example: consider the signal $x(t) = \sin \omega_0 t$.

$$\sin\omega_0 t = \frac{1}{2j}e^{j\omega_0 t} - \frac{1}{2j}e^{-j\omega_0 t}$$

Comparing the right-hand sides of this equation and Eq. (3.32), we have

To determine the Fourier series coefficients for x(t), we use Eq. (3.33). Because of the symmetry of x(t) about t = 0, we choose $-T/2 \le t \le T/2$ as the interval over which the integration is performed, although any other interval of length *T* is valid the thus lead to the same result.

For k = 0,

$$a_0 = \frac{1}{T} \int_{-T_1}^{T_1} x(t) dt = \frac{1}{T} \int_{-T_1}^{T_1} dt = \frac{2T_1}{T},$$

(3.36)

For $k \neq 0$, we obtain

$$a_{k} = \frac{1}{T} \int_{-T_{1}}^{T_{1}} e^{-jk\omega_{0}t} dt = -\frac{1}{jk\omega_{0}T} e^{-jk\omega_{0}t} \Big|_{-T_{1}}^{T_{1}}$$

$$=\frac{2}{k\omega_0 T}\left[\frac{e^{jk\omega_0 T_1} - e^{-jk\omega_0 T_1}}{2j}\right]$$

$$=\frac{2\sin(k\omega_0T_1)}{k\omega_0T}=\frac{\sin(k\omega_0T_1)}{k\pi}$$

The above figure is a bar graph of the Fourier series coefficients for a fixed T_1 and several values of T. For this example, the coefficients are real, so they can be depicted with a single graph. For complex coefficients, two graphs corresponding to the real and imaginary parts or amplitude and phase of each coefficient, would be required.

Queries ...?

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